



# High Speed PWM Controller

## FEATURES

- Compatible with Voltage or Current-Mode Topologies
- Practical Operation @ Switching Frequencies to 1.0MHz
- 50ns Propagation Delay to Output
- High Current Totem Pole Output (1.5A peak)
- Wide Bandwidth Error Amplifier
- Fully Latched Logic with Double Pulse Suppression
- Pulse-by-Pulse Current Limiting
- Soft Start/Max. Duty Cycle Control
- Under-Voltage Lockout with Hysteresis
- Low Start Up Current (1.1mA)
- Trimmed Bandgap Reference (5.1V  $\pm$ 1%)

## DESCRIPTION

The UC1823 family of PWM control ICs is optimized for high frequency switched mode power supply applications. Particular care was given to minimizing propagation delays through the comparators and logic circuitry while maximizing bandwidth and slew rate of the error amplifier. This controller is designed for use in either current-mode or voltage-mode systems with the capability for input voltage feed-forward.

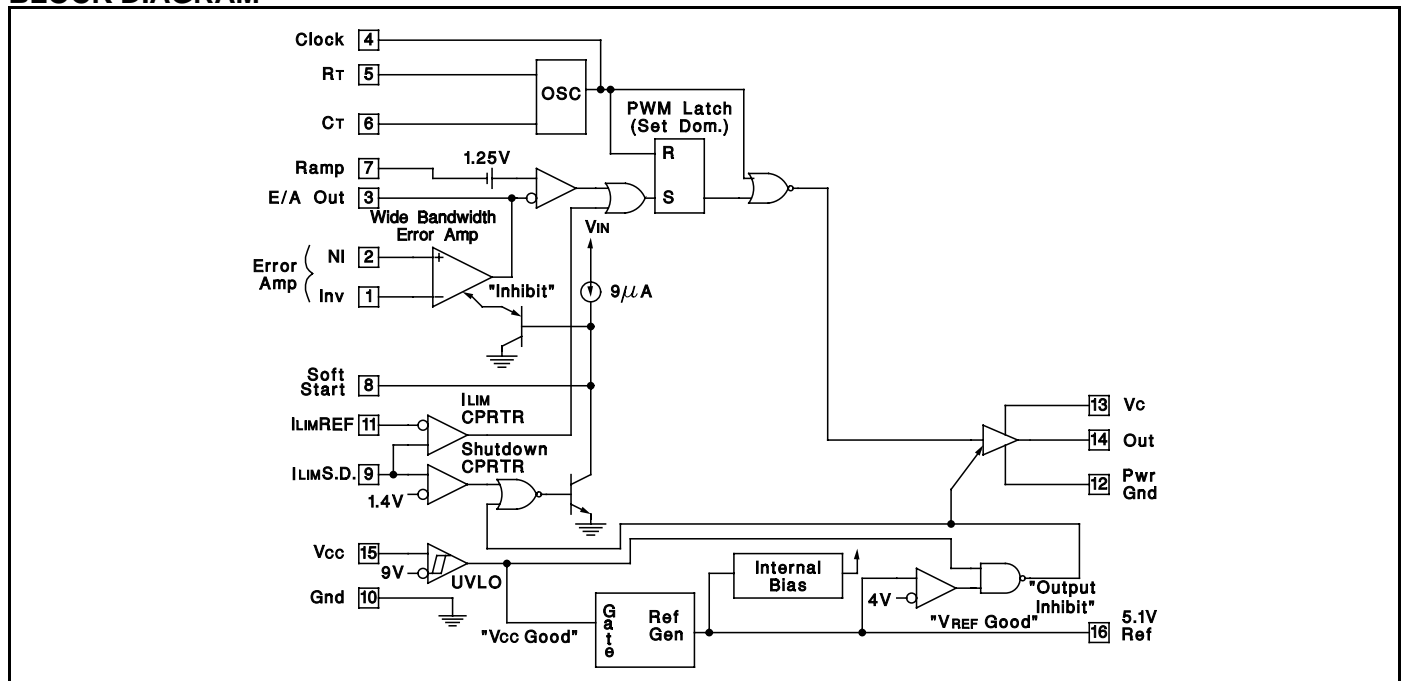
Protection circuitry includes a current limit comparator, a TTL compatible shutdown port, and a soft start pin which will double as a maximum duty cycle clamp. The logic is fully latched to provide jitter free operation and prohibit multiple pulses at the output. An under-voltage lockout section with 800mV of hysteresis assures low start up current. During under-voltage lockout, the output is high impedance. The current limit reference (pin 11) is a DC input voltage to the current limit comparator. Consult specifications for details.

These devices feature a totem pole output designed to source and sink high peak currents from capacitive loads, such as the gate of a power MOSFET. The on state is defined as a high level.

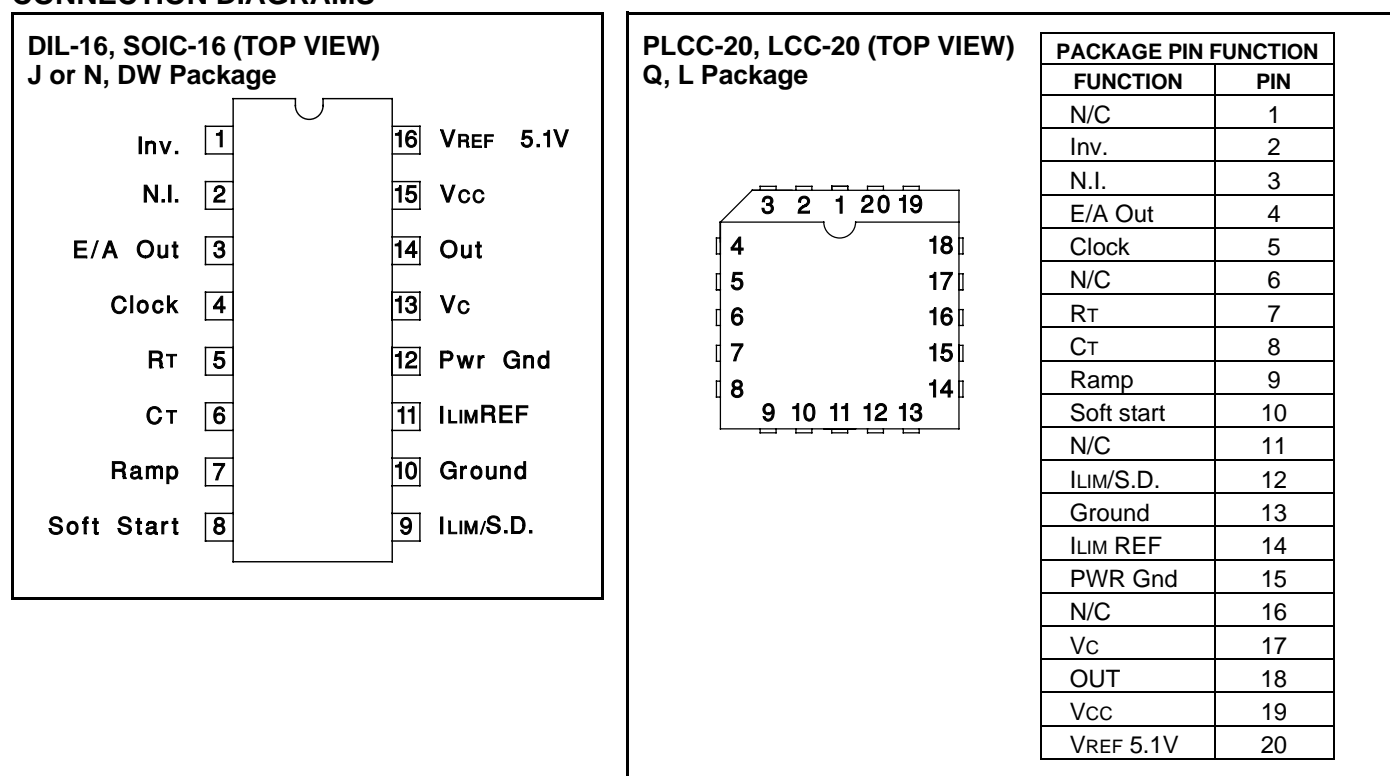
## ABSOLUTE MAXIMUM RATINGS

Supply Voltage (Pins 15, 13) .....	30V	Oscillator Charging Current (Pin 5) .....	-5mA
Output Current, Source or Sink (Pin14)		Power Dissipation at $T_A = 60^\circ\text{C}$ .....	1W
DC .....	0.5A	Storage Temperature Range .....	-65°C to +150°C
Pulse (0.5 $\mu$ s) .....	2.0A	Lead Temperature (Soldering, 10 seconds) .....	300°C
Analog Inputs (Pins 1, 2, 7, 8, 9, 11) .....	-0.3V to +6V	Note: All voltages are with respect to ground, Pin 10.	
Clock Output Current (Pin 4) .....	-5mA	Currents are positive into the specified terminal.	
Error Amplifier Output Current (Pin 3) .....	5mA	Consult Packaging Section of Databook for thermal limitations and considerations of packages.	
Soft Start Sink Current (Pin 8) .....	20mA		

## BLOCK DIAGRAM



## CONNECTION DIAGRAMS



**ELECTRICAL CHARACTERISTICS:** Unless otherwise noted, these specifications apply for  $R_T = 3.65k$ ,  $C_T = 1nF$ ,  $V_{CC} = 15V$ ,  $0^\circ C < T_A < +70^\circ C$  for the UC3823,  $-25^\circ C < T_A < +85^\circ C$  for the UC2823, and  $-55^\circ C < T_A < +125^\circ C$  for the UC1823,  $T_A = T_J$ .

PARAMETER	TEST CONDITIONS	UC1823 UC2823			UC3823			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
Reference Section								
Output Voltage	TJ = 25°C, lo = 1mA	5.05	5.10	5.15	5.00	5.10	5.20	V
Line Regulation	10 < Vcc < 30V		2	20		2	20	mV
Load Regulation	1 < lo < 10mA		5	20		5	20	mV
Temperature Stability*	TMIN < TA < TMAX		0.2	0.4		0.2	0.4	mV/°C
Total Output Variation*	Line, Load, Temp.	5.00		5.20	4.95		5.25	
Output Noise Voltage*	10Hz < f < 10kHz		50			50		μV
Long Term Stability*	TJ = 125°C, 1000 hrs.		5	25		5	25	mV
Short Circuit Current	VREF=0V	-15	-50	-100	-15	-50	-100	mA
Oscillator Section								
Initial Accuracy*	TJ=25°C	360	400	440	360	400	440	kHz
Voltage Stability*	10 < Vcc < 30V		0.2	2		0.2	2	%
Temperature Stability*	TMIN <TA < TMAX		5			5		%
Total Variation*	Line, Temp.	340		460	340		460	kHz
Clock Out High		3.9	4.5		3.9	4.5		V
Clock Out Low			2.3	2.9		2.3	2.9	V
Ramp Peak*		2.6	2.8	3.0	2.6	2.8	3.0	V
Ramp Valley*		0.7	1.0	1.25	0.7	1.0	1.25	V
Ramp Valley to Peak*		1.6	1.8	2.0	1.6	1.8	2.0	V

\* These parameters are guaranteed by design but not 100% tested in production.

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PARAMETER	TEST CONDITIONS	UC1823 UC2823			UC3823			UNITS
		MIN	TYP	MAX	MIN	TYP	MAX	
Error Amplifier Section								
Input Offset Voltage				10			15	mV
Input Bias Current			0.6	3		0.6	3	μA
Input Offset Current			0.1	1		0.1	1	μA
Open Loop Gain	1 < Vo < 4V	60	95		60	95		dB
CMRR	1.5 < VCM < 5.5V	75	95		75	95		dB
PSRR	10 < Vcc < 30V	85	110		85	110		dB
Output Sink Current	VPIN 3 =1V	1	2.5		1	2.5		mA
Output Source Current	VPIN 3 = 4V	-0.5	-1.3		-0.5	-1.3		mA
Output High Voltage	IPIN 3 = −0.5mA	4.0	4.7	5.0	4.0	4.7	5.0	V
Output Low Voltage	IPIN 3 = 1mA	0	0.5	1.0	0	0.5	1.0	V
Unity Gain Bandwidth*		3	5.5		3	5.5		MHz
Slew Rate*		6	12		6	12		V/μS
PWM Comparator Section								
Pin 7 Bias Current	VPIN 7 = 0V		-1	-5		-1	-5	μA
Duty Cycle Range		0		80	0		85	%
Pin 3 Zero D.C. Threshold	VPIN 7 = 0V	1.1	1.25		1.1	1.25		V
Delay to Output*			50	80		50	80	ns
Soft-Start Section								
Charge Current	VPIN 8 = 0.5V	3	9	20	3	9	20	μA
Discharge Current	VPIN 8 = 1V	1			1			mA
Current Limit/Shutdown Section								
Pin 9 Bias Current	0 < VPIN 9 < 4V			±10			±10	μA
Current Limit Offset	VPIN 11 = 1.1V			15			15	mV
Current Limit Common Mode Range (VPIN 11)		1.0		1.25	1.0		1.25	V
Shutdown Threshold		1.25	1.40	1.55	1.25	1.40	1.55	V
Delay to Output*			50	80		50	80	ns
Output Section								
Output Low Level	IOUT = 20mA		0.25	0.40		0.25	0.40	V
	IOUT = 200mA		1.2	2.2		1.2	2.2	V
Output High Level	IOUT = −20mA	13.0	13.5		13.0	13.5		V
	IOUT = −200mA	12.0	13.0		12.0	13.0		V
Collector Leakage	Vc = 30V		100	500		100	500	μA
Rise/Fall Time*	CL = 1nF		30	60		30	60	ns
Under-Voltage Lockout Section								
Start Threshold		8.8	9.2	9.6	8.8	9.2	9.6	V
UVLO Hysteresis		0.4	0.8	1.2	0.4	0.8	1.2	V
Supply Current								
Start Up Current	VCC = 8V		1.1	2.5		1.1	2.5	mA
Icc	VPIN 1, VPIN 7, VPIN 9 =0V, VPIN 2 = 1V		22	33		22	33	mA

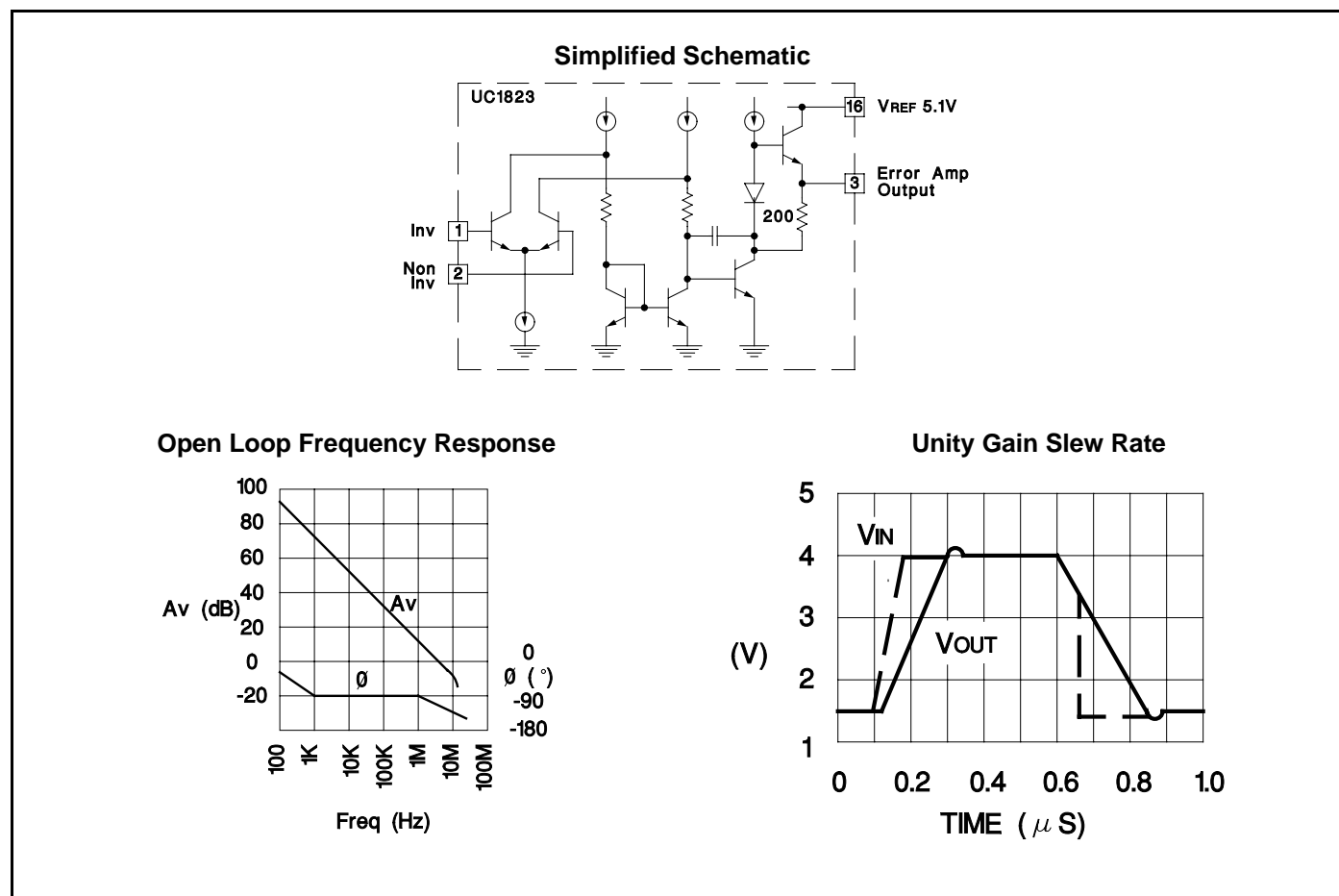
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## UC1823 PRINTED CIRCUIT BOARD LAYOUT CONSIDERATIONS

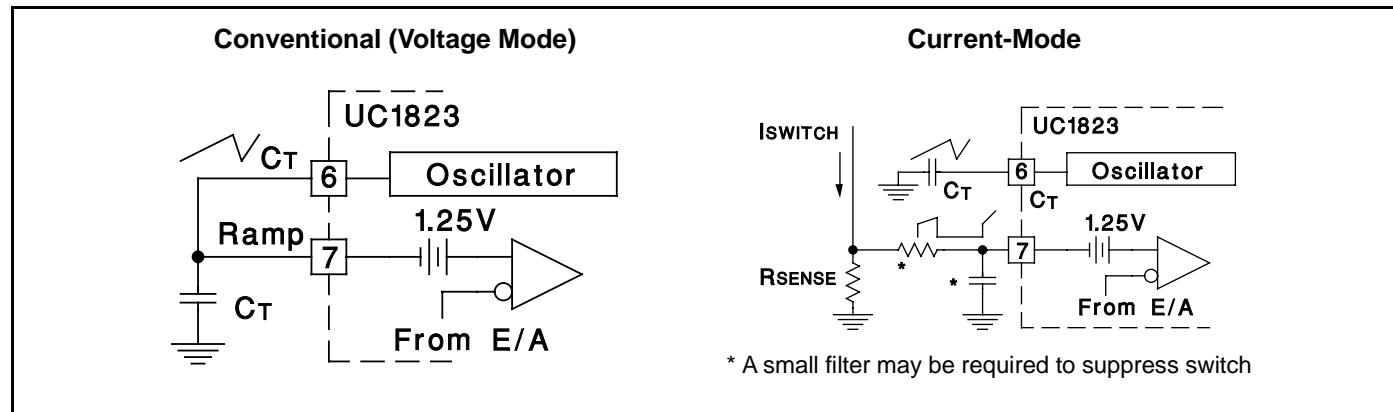
High speed circuits demand careful attention to layout and component placement. To assure proper performance of the UC1823, follow these rules. 1) Use a ground plane. 2) Damp or clamp parasitic inductive kick energy from the gate of driven MOSFET. Don't allow the output pins to ring below ground. A series gate resistor or a shunt 1 Amp Schottky diode at the output pin will serve

this purpose. 3) Bypass VCC, VC, and VREF. Use 0.1 $\mu$ F monolithic ceramic capacitors with low equivalent series inductance. Allow less than 1 cm of total lead length for each capacitor between the bypassed pin and the ground plane. 4) Treat the timing capacitor, CT, like a bypass capacitor.

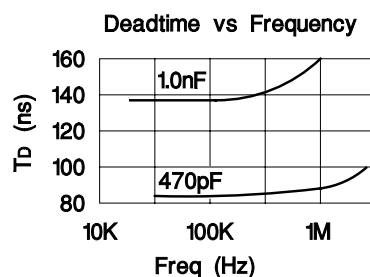
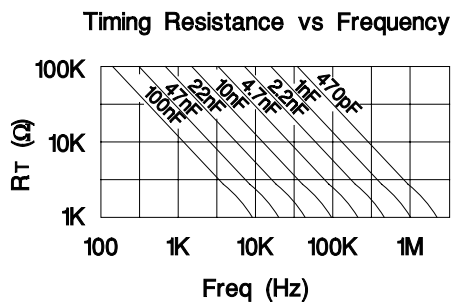
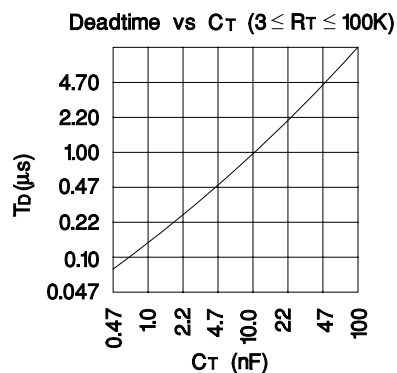
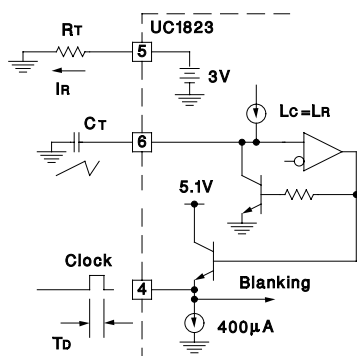
## ERROR AMPLIFIER CIRCUIT



## PWM APPLICATIONS

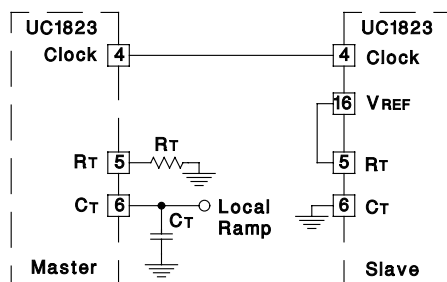


## OSCILLATOR CIRCUIT

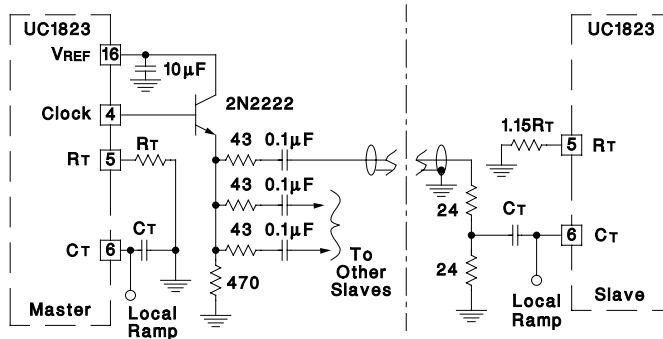


## SYNCHRONIZED OPERATION

### Two Units in Close Proximity

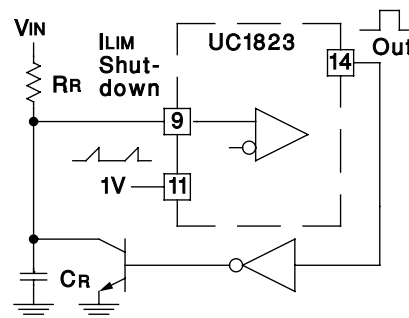


### Generalized Synchronization



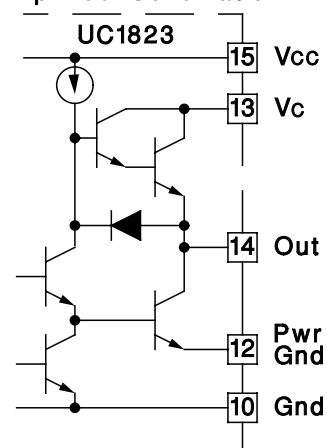
## CONSTANT VOLT-SECOND CLAMP CIRCUIT

The circuit shown here will achieve a constant volt-second product clamp over varying input voltages. The ramp generator components,  $R_T$  and  $C_R$  are chosen so that the ramp at Pin 9 crosses the 1V threshold at the same time the desired maximum volt-second product is reached. The delay through the inverter must be such that the ramp capacitor can be completely discharged during the minimum deadtime.

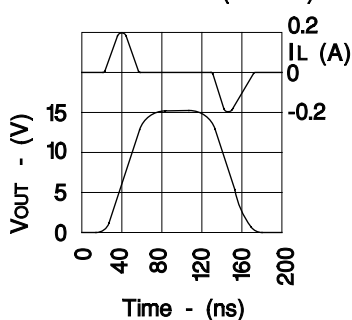


## OUTPUT SECTION

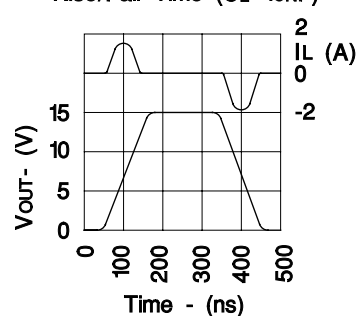
Simplified Schematic



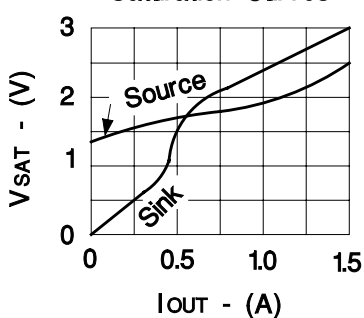
Rise/Fall Time ( $C_L=1nF$ )



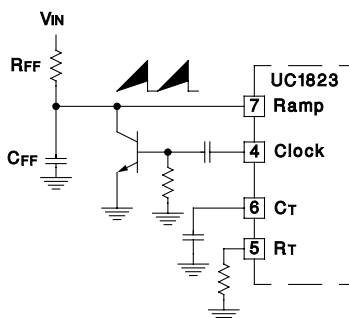
Rise/Fall Time ( $C_L=10nF$ )



Saturation Curves



## FEED FORWARD TECHNIQUE FOR OFF-LINE VOLTAGE MODE APPLICATION



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